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Determination of resonance frequencies of LC networks with binary link disorder

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Motivation

A LC circuit has a resonance frequency of $\omega_0=rac{1}{\sqrt{LC}}$

Frequency ω of AC current approaching the resonance frequency:

$$\lim_{\omega
ightarrow\omega_0}Z(\omega)=\infty$$

 \Rightarrow No power transmission possible!

Normal operation: ω should stay far below the smallest resonance.

Arbitrary LC network

- Coupled LC oscillators
- Set of resonance frequencies ω_n



Z = R + iX

$$rac{1}{X} = rac{1}{i\omega L} \ + i\omega C$$

Flow equations I

Combine Ohm's law

$$V_{ij}=Z_{ij}I_{ij} \quad ext{or} \quad I_{ij}=Y_{ij}V_{ij}$$

with Kirchhoff's laws for each node and mesh

$$I_i = \sum_j I_{ij}$$
 $V_{ij} = V_i - V_j$

to derive the current flow equations

$$I_i = \sum_j Y_{ij} (V_i - V_j)$$

Name	Symbol
Impedance matrix	\mathbf{Z}
Admittance matrix	Y



An arbitrary one-phasic AC grid with line impedances Z_{ij} .

 $\begin{array}{l} { { { Generator:}} I_i > 0 } \\ { { Consumer:} I_i < 0 } \end{array}$

 $\mathbf{Y} = \mathbf{Z}^{-1}$

R. Huang, G. Korniss, S. Nayak, 2009 (http://dx.doi.org/10.1103 /PhysRevE.80.045101)

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Flow Equations II

$$I_i = \sum_j Y_{ij} (V_i - V_j)$$

To get a standard matrix-vector multiplication, reformulate to

$$I_i = \sum_j L_{ij} V_j \quad ext{or} \quad \mathbf{I} = \mathbf{L} \mathbf{V}$$



An arbitrary one-phasic AC grid.

Generator: $I_i > 0$ Consumer: $I_i < 0$

with

$$L_{ij} = \delta_{ij} \sum_{k
eq i} Y_{ik} - (1 - \delta_{ij}) Y_{ij}$$

Note:

- \mathbf{L} is defined in analogy to the topological network Laplacian \mathbf{G} .
- L is commonly referred to as *admittance matrix* as well.

Basic Graph Theory I

A graph is given by

- a set of nodes *i*
- a set of edges (i, j)

Properties

Property	Explanation	
Order N	Number of nodes	
Size	Number of edges	





N=4

Basic Graph Theory II

Node Properties

Property	Explanation
Degree	Number of incident edges

Matrices

Name	Explanation
Degree matrix ${f D}$	Diagonal matrix containing all node degrees
Adjacency matrix ${f E}$	Nonzero only if nodes i and j are <i>adjacent</i>
Laplacian matrix ${f G}$	$\mathbf{G} = \mathbf{D} - \mathbf{E}$

$$G_{ij} = \delta_{ij} \sum_{k
eq i} E_{ik} - (1-\delta_{ij}) E_{ij}$$

Example



$$D = egin{pmatrix} 2 & 0 & 0 & 0 \ 0 & 2 & 0 & 0 \ 0 & 0 & 3 & 0 \ 0 & 0 & 0 & 1 \end{pmatrix} \ E = egin{pmatrix} 0 & 1 & 1 & 0 \ 1 & 0 & 1 & 0 \ 1 & 1 & 0 & 1 \ 0 & 0 & 1 & 0 \end{pmatrix} \ G = egin{pmatrix} 2 & -1 & 0 \ -1 & 2 & -1 & 0 \ -1 & 2 & -1 & 0 \ -1 & -1 & 3 & -1 \ 0 & 0 & -1 & 1 \end{pmatrix}$$

Basic Graph Theory III

Laplacian spectrum

Certain eigenvalues (EVs) of the Laplacian matrix ${f G}$ have special properties:

- All EVs are non-negative (as **G** is *positive semi-definite*).
- At least one eigenvalue is 0.
- Number of eigenvalues equal to 0: Number of *connected subgraphs*.
- Second-smallest EV: Algebraic connectivity.

Example







The regular 2D grid graph



N=9, periodic boundary conditions

The Small World Model

Closed ring with N nodes, plus S randomly chosen *shortcuts*.

Shortcut density
$$\frac{p}{2} = \frac{S}{N}$$

Number of shortcuts $S=rac{Np}{2}$

Two limiting cases:

Limiting case	Resulting graph	
$p_{\min}=0$	Closed ring	
$p_{ m max}=N-3$	Complete graph	

Example



Maximum number of shortcuts:

$$N=12 \qquad p=1.5 \ S_{
m max}=rac{N(N-3)}{2}$$

J. Travers, S. Milgram, 1969 (http://www.jstor.org/stable/2786545); M. Newman, D. Watts, 1999 (http://dx.doi.org/10.1016/S0375-9601%2899%2900757-4); R. Huang, G. Korniss, S. Nayak, 2009 (http://dx.doi.org/10.1103/PhysRevE.80.045101)

Adding Edge Attributes

In order to describe LC networks, we attribute an impedance Z_{ij} to each edge.

Here, we consider *random* impedances, using a *binary distribution*:

Edge type	$oldsymbol{Z}_{ij}$	Chance
Capacitance	$(i\omega C)^{-1}$	q
Inductance	$i\omega L$	1-q

R. Huang, G. Korniss, S. Nayak, 2009 (http://dx.doi.org/10.1103 /PhysRevE.80.045101) Example



$$N=12 \ p=1.5 \quad q=0.5$$

Edge type	Z_{ij}	Chance	Y_{ij}	$m{h}_{ij}$
Capacitance	$(i\omega C)^{-1}$	q	$y_1=i\omega C$	-1
Inductance	$i\omega L$	1-q	$y_2=(i\omega L)^{-1}$	1

Introduce matrix \mathbf{h} :

$$h_{ij} = \left\{egin{array}{ccc} -1 &, & ext{edge} \ (i,j) ext{ carries capacitance C} \ 1 &, & ext{edge} \ (i,j) ext{ carries inductance L} \ 0 &, & ext{no edge between i and j.} \end{array}
ight.$$

 \Rightarrow Similar to ${f E}$, but also -1 allowed.

Resonance frequencies I

Remember:

Flow equations:

$$egin{aligned} I_i = \sum_j L_{ij} V_j & ext{or} \ \mathbf{I} = \mathbf{L} \mathbf{V} \end{aligned}$$

with "Laplacian matrix" **L**.

- Consider resonance case, $I_i = 0$
- The system can be seen as a system of coupled LC oscillator circuits
- The system has N resonance frequencies ω_n

$$\sum_j L_{ij}(\omega) \, V_j = 0 \qquad ext{or} \qquad \mathbf{LV} = \mathbf{0}$$

Resonance frequencies II

Define \mathbf{H} ,

$$H_{ij} = \delta_{ij} \sum_{k
eq i} h_{ik} - (1-\delta_{ij}) h_{ij} \quad ,$$

Remember:
$$u = i w C$$

$$egin{aligned} y_1 &= i\omega 0 \ y_2 &= (i\omega L)^{-1} \ h_{ij} &= egin{aligned} &-1 &, & \mathrm{Y}_{\mathrm{ij}} &= \mathrm{y}_1 \ 1 &, & \mathrm{Y}_{\mathrm{ij}} &= \mathrm{y}_2 \ 0 &, & \mathrm{Y}_{\mathrm{ij}} &= 0 \end{aligned}$$

and

$$\lambda = rac{y_1+y_2}{y_1-y_2} \quad,$$

so that the flow equations for the resonance case can be rewritten as

$$\mathbf{LV} = \mathbf{0} \qquad \Rightarrow \qquad (\mathbf{H} - \lambda \mathbf{G})\mathbf{V} = \mathbf{0}$$

But this is not yet a regular eigenvalue problem...

R. Huang, G. Korniss, S. Nayak, 2009 (http://dx.doi.org/10.1103 /PhysRevE.80.045101); Fyodorov 1999 (http://dx.doi.org/10.1088/0305-4470 /32/42/314)

Resonance frequencies III

Define

Remember:

$$ilde{\mathbf{H}} = \mathbf{G}^{-1/2} \, \mathbf{H} \, \mathbf{G}^{-1/2}$$

(real symmetric) and

$$egin{aligned} y_1 &= i\omega C \ y_2 &= (i\omega L)^{-1} \end{aligned}$$

 $\lambda=rac{y_1+y_2}{y_1-y_2}$

 $ilde{\mathbf{V}} = \mathbf{G}^{1/2}\mathbf{V}$

Notes:

- 1. G is real and positive semi-definite, so its matrix square root $G^{1/2}$ is uniquely defined.
- 2. G is positive semi-definite, i.e. it has at least one eigenvalue 0 and hence is always singular. So its pseudo-inverse G^{-1} has to be considered.

Fyodorov 1999 (http://dx.doi.org/10.1088/0305-4470/32/42/314)

Resonance frequencies IV

Define

Remember: $\lambda = rac{y_1+y_2}{y_1-y_2}$

$$ilde{\mathbf{H}} = \mathbf{G}^{-1/2} \, \mathbf{H} \, \mathbf{G}^{-1/2}$$

(real symmetric) and

$$egin{aligned} y_1 &= i\omega C \ y_2 &= (i\omega L)^{-1} \end{aligned}$$

$$ilde{\mathbf{V}} = \mathbf{G}^{1/2} \mathbf{V}$$

Then, we can rewrite

$$(\mathbf{H} - \lambda \mathbf{G}) \mathbf{V} = \mathbf{0} \qquad \Rightarrow \qquad \mathbf{ ilde{H}} \, \mathbf{ ilde{V}}_n = \lambda_n \mathbf{ ilde{V}}_n$$

So we are facing a regular eigenvalue problem, with a known relationship between the eivenvalues λ_n

and the resonance frequencies ω_n :

$$\omega_n = rac{1}{\sqrt{LC}} \, \sqrt{rac{1+\lambda_n}{1-\lambda_n}}$$

Fyodorov 1999 (http://dx.doi.org/10.1088/0305-4470/32/42/314)

Resonance spectra I: 2D grid

Define the density of resonances (DOR):

$$ho(\lambda) = rac{1}{N}\sum_{n=1}^{N_{
m R}} \delta(\lambda-\lambda_n)$$

 $N_{
m R}$: Number of "true" resonances $(-1 < \lambda_n < 1)$

Obtain *ensemble average* (arithmetic mean) of the DOR over many disorder realizations (ADOR).



Resonance spectra II: Small World Model

Large-p limit



 \Rightarrow Confirming results by Huang et al. (http://dx.doi.org/10.1103 /PhysRevE.80.045101)

Resonance spectra III: Small World Model

Small-p limit



 \Rightarrow Largely confirming results by Huang et al. (http://dx.doi.org/10.1103 /PhysRevE.80.045101)

 \Rightarrow Difference: Peak at $\lambda = 0$.

Summary & Outlook

Summary

- Description of LC networks
 - simple graphs
 - binary distribution of edge impedances
- Calculation of resonance frequencies and the density of resonances

Outlook

- Different topologies (triangular grid, honeycomb grid, and **realistic network topologies**).
- Other impedance distributions (also **continuous distributions**), also including **ohmic resistances**.
- Beyond the resonance case (current and power flow calculations).

Thank you for your attention!

References

- J. Travers, S. Milgram, 1969 (http://www.jstor.org/stable/2786545)
- M. Newman, D. Watts, 1999 (http://dx.doi.org/10.1016 /S0375-9601%2899%2900757-4)
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